# **Digital Signal Processing**

#### **Periodic Sampling**

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# **Periodic Sampling**

- Periodic sampling
  - Process of representing a continuous signal with a sequence of discrete data values
  - In practice, sampling is performed by applying a continuous signal to an analog-to-digital (A/D) converter
  - Primary concern is how fast a given continuous signal must be sampled to preserve its information content

#### Example: given following sequence of values

x(0) = 0, x(1) = 0.866, x(2) = 0.866, x(3) = 0, x(4) = -0.866, x(5) = -0.866, x(6) = 0

- They represent values of a time-domain sinewave taken at periodic intervals
- Draw that sinewave

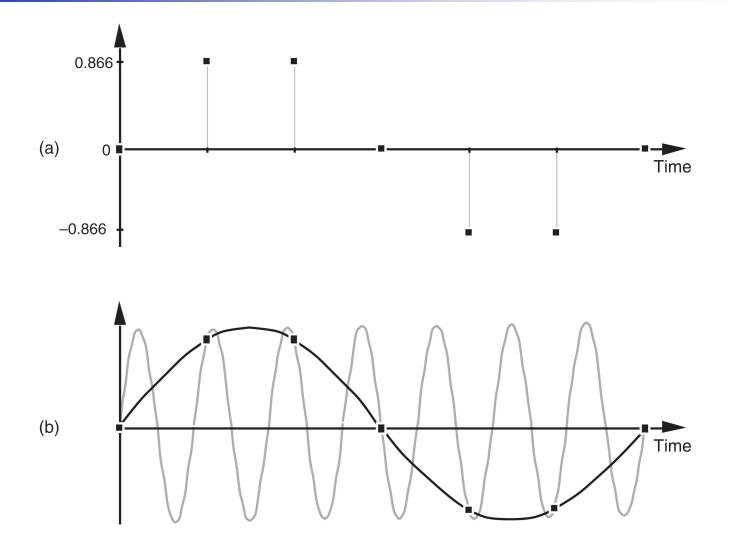


Figure 2-1 Frequency ambiguity: (a) discrete-time sequence of values; (b) two different sinewaves that pass through the points of the discrete sequence.

#### Frequency ambiguity

 If data sequence represents periodic samples of a sinewave, we cannot unambiguously determine frequency of sinewave from those sample values alone

#### Mathematical origin of frequency ambiguity

 $x(t) = \sin(2\pi f_o t)$ 

 $x(n) = \sin(2\pi f_o n t_s) = \sin(2\pi f_o n t_s + 2\pi m) = \sin(2\pi (f_o + \frac{m}{n t_s}) n t_s)$ 

$$\xrightarrow{\text{if } m=kn} x(n) = \sin(2\pi(f_o + \frac{k}{t_s})nt_s)$$
$$\xrightarrow{f_s=1/t_s} x(n) = \sin(2\pi f_o nt_s) = \sin(2\pi(f_o + kf_s)nt_s)$$

When sampling at a rate of f<sub>s</sub> samples/second, if k is any positive or negative integer, we cannot distinguish between sampled values of a sinewave of f<sub>o</sub> Hz and a sinewave of (f<sub>o</sub>+kf<sub>s</sub>) Hz

- Frequency ambiguity (*aliasing*) effects
  - Spectrum of any discrete series of sampled values contains periodic replications of original continuous spectrum
  - Period between these replicated spectra in frequency domain is always f<sub>s</sub>
  - Spectral replications repeat all the way in both directions of frequency spectrum

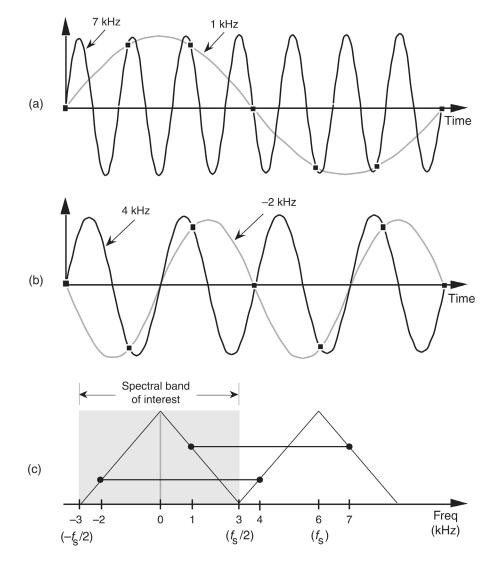


Figure 2-2 Frequency ambiguity effects of Eq. (2-5): (a) sampling a 7 kHz sinewave at a sample rate of 6 kHz; (b) sampling a 4 kHz sinewave at a sample rate of 6 kHz; (c) spectral relationships showing aliasing of the 7 and 4 kHz sinewaves.

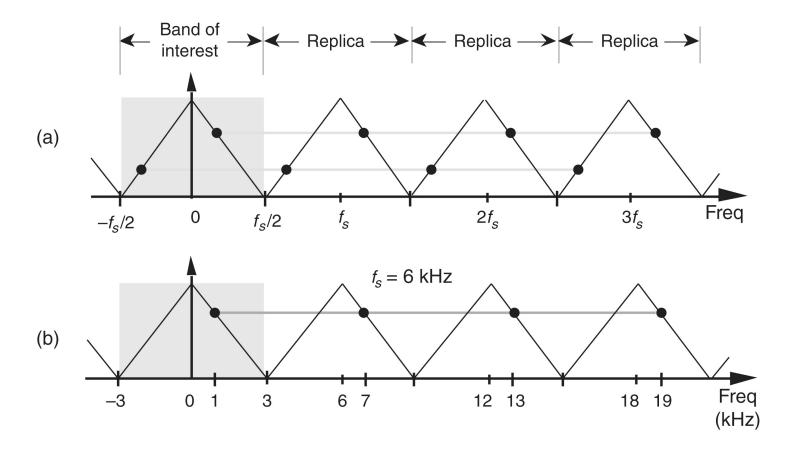
- Fig. 2-2(a)
  - $f_0 = 7 \text{ kHz}, f_s = 6 \text{ kHz}$
  - $k = -1 \rightarrow f_0 + kf_s = [7 + (-1 \cdot 6)] = 1 \text{ kHz}$
  - No processing scheme can determine if sequence of sampled values came from a 7 kHz or a 1 kHz sinusoid
  - 1 kHz is an alias of 7 kHz

Fig. 2-2(b)

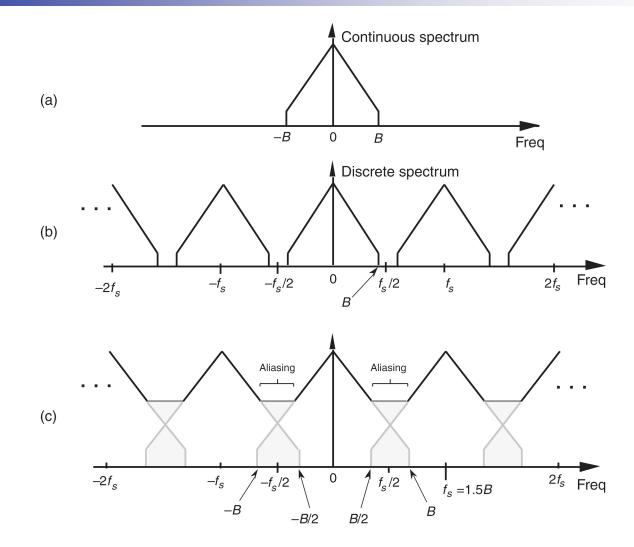
- $f_0 = 4 \text{ kHz}, f_s = 6 \text{ kHz}$
- $k = -1 \rightarrow f_0 + kf_s = [4 + (-1 \cdot 6)] = -2 \text{ kHz}$

### Fig. 2-2(c)

- *f<sub>s</sub>*/2 is an important quantity, referred to by critical Nyquist, half Nyquist, or folding frequency
- We're interested in signal components that are aliased into frequency band between  $-f_s/2$  and  $+f_s/2$



**Figure 2-3** Shark's tooth pattern: (a) aliasing at multiples of the sampling frequency; (b) aliasing of the 7 kHz sinewave to 1 kHz, 13 kHz, and 19 kHz.



**Figure 2-4** Spectral replications: (a) original continuous lowpass signal spectrum; (b) spectral replications of the sampled lowpass signal when  $f_s/2 > B$ ; (c) frequency overlap and aliasing when the sampling rate is too low because  $f_s/2 < B$ .

### Fig. 2-4(a)

- Spectrum of a continuous real-valued lowpass x(t) signal
- Spectrum is symmetrical around zero Hz
- Signal is band-limited
  - Its spectral amplitude is zero above +B Hz and below
    B Hz
- x(t) time signal is called a *lowpass signal* because its spectral energy is low in frequency
- Spectrum of a continuous signal *cannot* be represented in a digital machine in its current band-limited form → replicated form of (b)

### Nyquist criterion

- *f<sub>s</sub>* ≥ 2*B*, to separate spectral replications at *folding* frequencies of ±*f<sub>s</sub>*/2
- Fig. 2-4(c)
  - Sampling frequency is lowered to  $f_s = 1.5B$  Hz
  - Lower edge and upper edge of spectral replications centered at +f<sub>s</sub> and -f<sub>s</sub> now lie in band of interest
    - Equivalent to original spectrum folding to left at  $+f_s/2$  and folding to right at  $-f_s/2$
    - Spectral information in bands of -B to -B/2 and B/2 to B Hz is corrupted (aliasing errors)

- A key property of band  $\pm f_s/2$  Hz
  - Entire spectral content (any signal energy located above +B Hz and below -B Hz) of original continuous spectrum always ends up in band of interest between  $-f_s/2$  and  $+f_s/2$  after sampling, regardless of sample rate
  - For this reason, continuous (analog) *lowpass* filters are necessary in practice

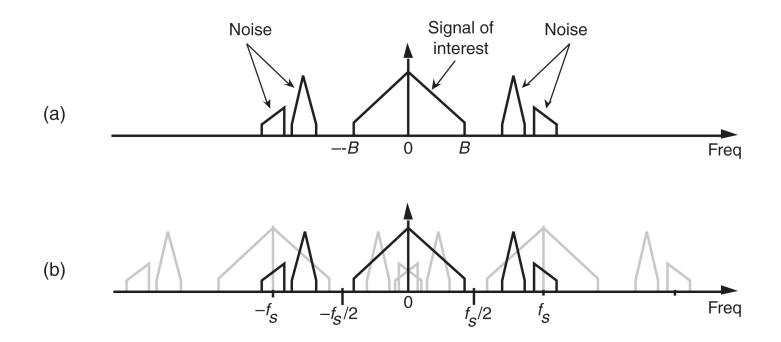
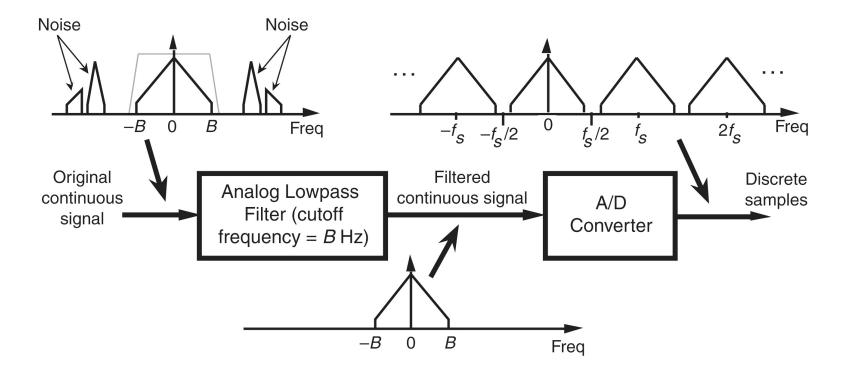


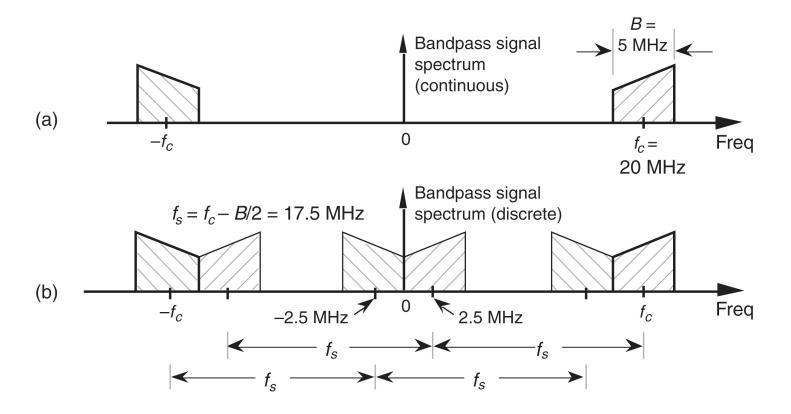
Figure 2-5 Spectral replications: (a) original continuous signal-plus-noise spectrum; (b) discrete spectrum with noise contaminating the signal of interest.



**Figure 2-6** Lowpass analog filtering prior to sampling at a rate of  $f_s$  Hz.

#### Bandpass sampling

- A technique to sample a continuous bandpass signal that is centered about some frequency other than zero Hz
- Reduces speed requirement of A/D converters below that necessary with traditional lowpass sampling
- Reduces amount of digital memory necessary to capture a given time interval of a continuous signal
- We're more concerned with a signal's bandwidth than its highest-frequency component



**Figure 2-7** Bandpass signal sampling: (a) original continuous signal spectrum; (b) sampled signal spectrum replications when sample rate is 17.5 MHz.

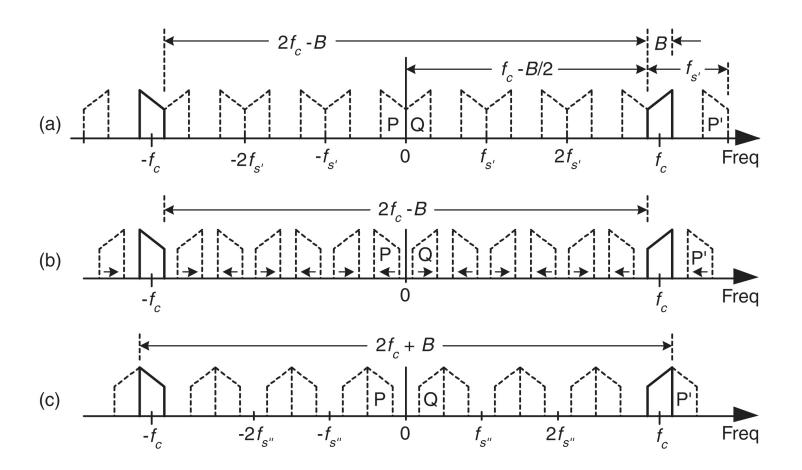
### Fig. 2-7(a)

- Negative frequency portion of signal is mirror image of positive frequency portion (real signal)
- Highest-frequency = 22.5 MHz
- Nyquist criterion → sampling frequency must be a minimum of 45 MHz

### Fig. 2-7(b)

- If sample rate is 17.5 MHz, spectral replications are located exactly at baseband
- Sampling at 45 MHz was unnecessary to avoid aliasing—instead we've used spectral replicating effects to our advantage

- Sampling translation
  - Bandpass sampling performs digitization and frequency translation in a single process
- We can sample at some still lower rate and avoid aliasing



**Figure 2-8** Bandpass sampling frequency limits: (a) sample rate  $f_{s'} = (2f_c - B)/6$ ; (b) sample rate is less than  $f_{s'}$ ; (c) minimum sample rate  $f_{s''} < f_{s'}$ .

### Fig. 2-8(a)

- Continuous input bandpass signal of bandwidth B
- Carrier frequency (signal is centered at) =  $f_c$  Hz
- Sample rate = f<sub>s'</sub> Hz → spectral replications of positive and negative bands, Q and P, butt up against each other at zero Hz

$$mf_{s'} = 2f_c - B$$
 or  $f_{s'} = \frac{2f_c - B}{m}$ 

- *m* = an arbitrary number of replications in the range of 2*f<sub>c</sub> B*
  - *m* can be any positive integer so long as *f<sub>s'</sub>* is never less than 2*B*

### Fig. 2-8

- If f<sub>s'</sub> is increased, original spectra (bold) do not shift, but all replications will shift
- At zero Hz, P band shifts to right, and Q band shifts to left
- These replications will overlap and aliasing occurs
- Thus, for an arbitrary *m*, there is a frequency that sample rate must not exceed

$$f_{s'} \le \frac{2f_c - B}{m}$$

#### Fig. 2-8(b) and (c)

- If we reduce sample rate below f<sub>s'</sub> shown in (a), spacing between replications will decrease in direction of arrows in (b)
  - Original spectra do not shift
- At some sample rate  $f_{s''}$  ( $f_{s''} < f_{s'}$ ), replication P' will butt up against positive original spectrum at  $f_c$  as shown in (c)

$$(m+1)f_{s''} = 2f_c + B$$
 or  $f_{s''} = \frac{2f_c + B}{m+1}$ 

•  $f_{s''}$  decreased  $\rightarrow$  aliasing occurs

$$f_{s"} \ge \frac{2f_c + B}{m+1}$$

#### To avoid aliasing, f<sub>s</sub> may be chosen anywhere in the range

$$\frac{2f_c - B}{m} \ge f_s \ge \frac{2f_c + B}{m+1} \tag{1}$$

• *m* is an arbitrary, positive integer ensuring  $f_s \ge 2B$ 

- Example (Fig. 2-7(a))
  - $f_c = 20 \text{ MHz}, B = 5 \text{ MHz}$

m	(2f <sub>c</sub> -B) / m	(2f <sub>c</sub> +B) / (m+1)	Optimum sampling rate
1	35.0 MHz	22.5 MHz	22.5 MHz
2	17.5 MHz	15.0 MHz	17.5 MHz
3	11.66 MHz	11.25 MHz	11.25 MHz
4	8.75 MHz	9.0 MHz	
5	7.0 MHz	7.5 MHz	

Sample rates below 11.25 MHz unacceptable

• Will not satisfy Eq. (1) as well as  $f_s \ge 2B$ 

 Optimum sampling frequency is the frequency where spectral replications butt up against each other at zero Hz

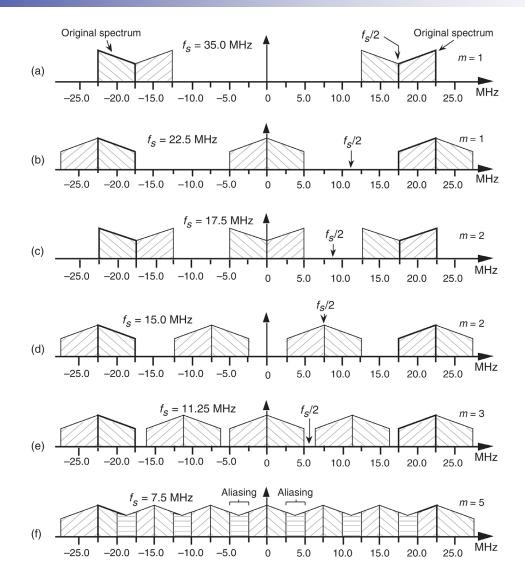
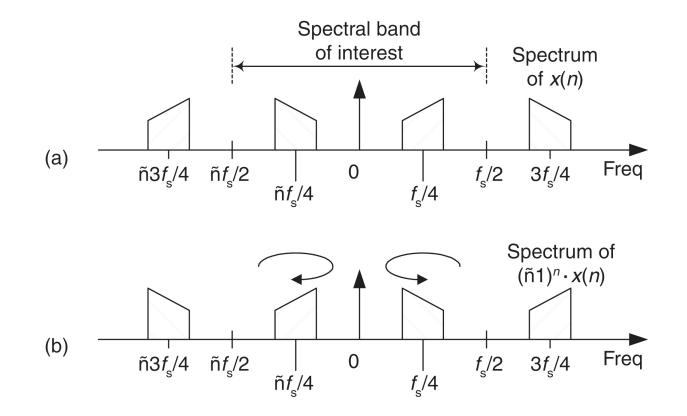


Figure 2-9 Various spectral replications from Table 2-1: (a)  $f_s = 35$  MHz; (b)  $f_s = 22.5$  MHz; (c)  $f_s = 17.5$  MHz; (d)  $f_s = 15$  MHz; (e)  $f_s = 11.25$  MHz; (f)  $f_s = 7.5$  MHz.

- Spectral Inversion in Bandpass Sampling
  - Some of permissible f<sub>s</sub> values from Eq. (1) provide a sampled baseband spectrum (located near zero Hz) that is inverted from original analog signal's positive and negative spectral shapes

Happens when m, in Eq. (1), is an odd integer

- We can invert spectrum back to its original orientation
- Discrete spectrum of any digital signal can be inverted by multiplying signal's discrete-time samples by (-1)<sup>n</sup>
- Center of flipping is  $f_s/4$  Hz (and  $-f_s/4$  Hz)
- When original positive spectral bandpass components are symmetrical about f<sub>c</sub> frequency, spectral inversion presents no problem



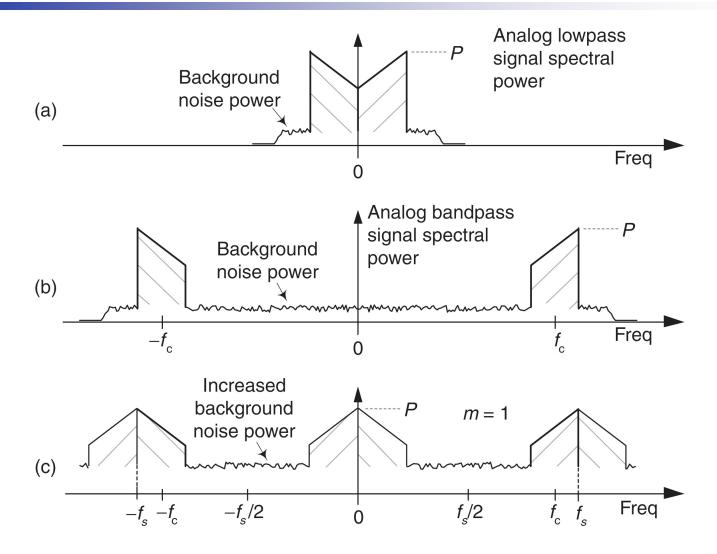
**Figure 2-10** Spectral inversion through multiplication by  $(-1)^n$ : (a) spectrum of original x(n); (b) spectrum of  $(-1)^n \cdot x(n)$ .

- Positioning sampled spectra at  $f_s/4$ 
  - In many signal processing applications it is useful to use an f<sub>s</sub> bandpass sampling rate that forces sampled spectra to be centered exactly at ±f<sub>s</sub>/4
  - To ensure that sampled spectra reside at ±f<sub>s</sub>/4, select f<sub>s</sub> using

$$f_s = \frac{4f_c}{2k-1}$$
, where  $k = 1, 2, 3, ...$ 

- Noise in bandpass-sampled signals
  - Signal-to-noise ratio (SNR) is ratio of power of a signal over total background noise power
  - Negative aspect of bandpass sampling
    - SNR of digitized signal is degraded
    - All of background spectral noise (Fig. 2-11(b)) resides in range of  $-f_s/2$  to  $f_s/2$  (Fig. 2-11(c))
    - Bandpass-sampled background noise power increases by a factor of *m* + 1 (denominator of right-side ratio in Eq. (1)) while signal power *P* remains unchanged
    - Bandpass-sampled signal's SNR is reduced by

 $D_{SNR} = 10 \cdot \log_{10}(m+1) dB$ below SNR of original analog signal



**Figure 2-11** Sampling SNR degradation: (a) analog lowpass signal spectral power; (b) analog bandpass signal spectral power; (c) bandpass-sampled signal spectral power when m = 1.

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