Design of Digital Systems II Sequential-Circuit Design with Verilog

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- Majority of Verilog-based digital design is directed to clocked, synchronous systems that use edge-triggered flip-flops
- Like combinational behavior, edge-triggered behavior is specified using always blocks
 - Difference between combinational and edge-triggered behavior is in sensitivity list of always block
 - Keyword posedge or negedge is placed in front of signal name to indicate that the statements in block should be executed only at positive (rising) or negative (falling) edge of named signal
 - Table 1: Behavioral Verilog for a positive-edge-triggered D flip-flop.

```
module VrposDff(CLK, D, Q);
input CLK, D;
output Q;
reg Q;
always @ (posedge CLK)
Q <= D;
endmodule
```

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Table 2: Behavioral Verilog for a positive-edge-triggered D flip-flop with preset and clear.

```
module Vrposdffpc(CLK, PR_L, CLR_L, D, Q);
input CLK, PR_L, CLR_L, D;
output Q;
reg Q;
always @ (posedge CLK or negedge PR_L or negedge CLR_L)
if (PR_L==0) Q<= 1;
else if (CLR_L==0) Q <= 0;
else Q <= D;
endmodule
```

- Tab. 2 models a positive edge-triggered D flip-flop with asynchronous active-low preset and clear inputs
 - An edge-sensitivity keyword, negedge, is applied to a level, asynchronous input
 - Verilog compilers are set up to recognize this particular representation of edge-triggered-plus-asynchronous behavior, and in synthesis they will pick up right flip-flop component to implement it

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Table 3: Two modules for a positive-edge-triggered D ff with a QN output.

```
module VrposDffQN1(CLK, D, Q, QN);
  input CLK, D;
  output Q, QN;
  reg Q, QN;
  always @ (posedge CLK) begin
    Q \leq D; QN \leq !D;
  end
endmodule
module VrposDffQN2(CLK, D, Q, QN);
  input CLK, D;
  output Q, QN;
  reg Q, QN;
  always @ (posedge CLK)
    Q <= D;
  always @ (Q)
    QN <= !Q:
endmodule
```

• Tab. 3

- A typical synthesis tool infers two separate D flip-flops from the first module—one for Q and the other for QN
- $\, \bullet \,$ For the second module, QN is generated from Q using an inverter

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- Always use nonblocking assignments (<=) in sequential always blocks
 - In programs with multiple sequential always blocks using *blocking* assignments (=), simulation results can vary depending on the order in which the simulator chooses to execute those blocks
 - Using *nonblocking* assignments ensures that righthand sides of all assignments are evaluated before new values are assigned to any of lefthand sides
 - This makes results independent of order in which righthand sides are evaluated

Table 4: Clock generation within a test bench.

```
`timescale 1 ns / 1 ns
module mclkgen(MCLK);
   output MCLK;
   reg MCLK;
initial begin
   MCLK = 1; // Start clock at 1 at time 0
end
always begin // 10 ns clock generation
  #6 MCLK = 0; // 6 ns HIGH
  #4 MCLK = 1; end; // 4 ns LOW
endmodule
```

• Tab. 4 shows generation of a 100-MHz clock with a 60% duty cycle

- At time 0, MCLK is set to 1 by initial block
- Then, always block waits 6 ns, sets MCLK to 0, waits 4 ns, sets MCLK to 1, and repeats forever
- `timescale directive is used to set up the simulator both a default time unit and a precision of 1 ns

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State-Machine Design with Verilog

- There are many possible coding styles for creating state machines in Verilog, including using no consistent style at all
 - Without discipline of a consistent coding style, it is easy to write syntactically correct code where simulator's operation, synthesized hardware's operation, and what we think the machine should be doing are all different

• In Verilog state-machine coding style, code is divided into three parts

- State memory
 - This can be specified in behavioral form using an always block that is sensitive to a signal edge
 - Or it can use a structural style with explicit flip-flop instantiations
- Next-state (excitation) logic
 - This is written as a combinational always block whose sensitivity list includes machine's current state and inputs
 - This block usually contains a case statement that enumerates all values of current state
- Output logic
 - This is another combinational always block that is sensitive to current state and inputs
 - It may or may not include a case statement, depending on complexity of output function

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State-Machine Design with Verilog



Figure 1: Moore state-machine structure implied by Verilog coding style.

- Detailed coding within each section may very
 - When there is a tight coupling of next-state and output-logic specifications, it may be desirable to combine them into a single combinational always block, and into a single case statement
 - When pipelined outputs are used, output memory could be specified along with state memory, or a separate process or structural code could be used

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- Design a clocked synchronous state machine with two inputs, A and B, and a single output Z that is 1 if
 - A had the same value at each of two previous clock ticks, or
 - B has been 1 since the last time that the first condition was true

Otherwise, output should be 0

• One approach to writing a program is to construct a state and output table by hand and then manually convert it into a corresponding program

	A B					
s	00	01	11	10	z	
INIT	A0	A0	A1	A1	0	
A0	OK0	OK0	A1	A1	0	
A1	A0	A0	OK1	OK1	0	
OK0	OK0	OK0	OK1	A1	1	
OK1	A0	OK0	OK1	OK1	1	
			S*			

Table 5: State and output table for the example state machine.

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Table 6: Verilog program for state-machine example.

```
module VrSMex( CLOCK, A, B, Z );
  input CLOCK, A, B;
 output Z;
 reg Z;
 reg [2:0] Sreg, Snext;
                                // State register and next state
  parameter [2:0] INIT = 3'b000, // Define the states
                  AO
                     = 3'b001,
                       = 3'b010.
                  A1
                  0K0 = 3'b011.
                  OK1 = 3'b100;
 always @ (posedge CLOCK) // Create the state memory
    Sreg <= Snext;
 always @ (A, B, Sreg) begin // Next-state logic
    case (Sreg)
      INIT:
              if (A==0) Snext = A0;
              else
                         Snext = A1:
      AQ:
              if (A==0) Snext = OKO;
              else
                         Snext = A1;
              if (A==0) Snext = A0;
      A1:
                         Snext = OK1;
      OKO:
              if (A==0) Snext = OKO:
              else if ((A==1) && (B==0)) Snext = A1;
              else
                                         Snext = OK1;
              if ((A==0) && (B==0))
      OK1:
                                         Snext = AO:
              else if ((A==0) && (B==1)) Snext = OKO;
              else
                                         Snext = OK1:
      default Snext = INIT:
    endcase
  end
 always @ (Sreg)
                        // Output logic
   case (Sreg)
      INIT, AO, A1: Z = 0;
                   Z = 1:
      OKO. OK1:
      default
                    Z = 0:
    endcase
endmodule
```

		A B				
s	00	01	11	10	z	
NIT	A0	A0	A1	A1	0	
A0	OK0	OK0	A1	A1	0	
A1	A0	A0	OK1	OK1	0	
OK0	OK0	OK0	OK1	A1	1	
OK1	A0	OK0	OK1	OK1	1	
			S*			

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- First always block (state memory creation) in Tab. 6
 - During synthesis, positive-edge-triggered D flip-flops are inferred for Sreg
 - A synchronous or asynchronous RESET signal is easily provided as shown in Tab. 7

Table 7: Synchronous and asynchronous reset for state machines in Verilog.

```
// State memory with active-high synchronous reset
always @ (posedge CLOCK) // Create state memory
    if (RESET==1) Sreg <= INIT; else Sreg <= Snext;</pre>
```

// State memory with active-high asynchronous reset
always @ (posedge CLOCK or posedge RESET) // Create state memory
 if (RESET==1) Sreg <= INIT; else Sreg <= Snext;</pre>

- Second always block (next-state logic) in Tab. 6
 - default case at the end handles unused states, but not uncovered input combinations in other cases
 - In each case, an if statement and a final else is used to ensure that a value is always assigned to Snext
 - If there were any state/input combinations in which no value was assigned to Snext, Verilog compiler would infer an unwanted latch for Snext
 - Variations
 - To establish a default next state for machine if case fails to cover all state/input combinations, precede case statement with "Snext = INIT"
 - When most transitions stay in current state, case statement can be preceded by "Snext = Sreg"
 - Preceding with "Snext = 3'bx" is another variation which is useful in simulation to detect unspecified state/input combinations

- Third always block (output logic) in Tab. 6
 - Handles machine's single Moore output, Z, which is set to a value as a function of current state
 - To define Mealy output, output should be a function of inputs as well as state in each enumerated case
 - Inputs should also be added to sensitivity list of always block

• Pipelined outputs

- In design of high-speed circuits, it is often necessary to ensure that state-machine outputs are available as early as possible and do not change during each clock period
- One way to get this behavior is to encode state so that state variables themselves serve as outputs
 - Called *output-coded state assignment*
 - It yields a Moore machine in which output logic is wires
- Another approach is to design state machine so that outputs during one clock period depend on state and inputs during *previous* clock period
 - Called pipelined outputs
 - Obtained by attaching another stage of memory (flip-flops) to a machine's outputs



Figure 2: Mealy machine with pipelined outputs.

- In our example Verilog state machine, the module defines a Moore-type state machine with the structure shown in Fig. 1
 - We can convert the machine to have pipelined outputs with the structure shown in Fig. 3
 - To do this, we need only to declare a "next-output" variable Zn and replace the original Verilog state-memory and output code with code shown in Tab. 8

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Table 8: Verilog pipelined output code.

always @ (posedge CLOCK) // Create output register -- this code Z <= Zn; // could be combined with state-memory code if desired

```
always @ (Snext) // Output logic
case (Snext)
INIT, A0, A1: Zn = 0;
OK0, OK1: Zn = 1;
default Zn = 0;
endcase
```

- New machine's behavior is indistinguishable from that of original machine, except for timing
 - We have reduced propagation delay from CLOCK to Z by producing Z directly on a register output
 - But we have increased setup-time requirements of A and B to CLOCK
 - In addition to their propagation delay through next-state logic, changes in A and B must also get through output logic in time to meet setup time requirement of output flip-flop's D input

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- Direct Verilog coding without a state table
 - It is possible to write a Verilog program directly, without writing out a state table by hand
 - Design a clocked synchronous state machine with two inputs, A and B, and a single output Z that is 1 if
 - A had the same value at each of two previous clock ticks, or
 - $\bullet\,$ B has been 1 since the last time that the first condition was true

Otherwise, output should be 0

- We need to have a register that keeps track of A (LASTA)
- Three states must be defined
 - INIT
 - LOOKING: still looking for a match
 - OK: got a match or B has been 1 since last match

Table 9: Simplified Verilog state-machine design.

```
module VrSMexa( CLOCK, RESET, A, B, Z );
  input CLOCK, RESET, A, B;
  output Z;
  wire Z:
                      // declared as wire for continuous assignment
 reg LASTA;
                                   // LASTA holds last value of A
 reg [1:0] Sreg, Snext;
                                  // State register and next state
  parameter [1:0] INIT = 2'b00, // Define the states
                  LOOKING = 2'b10,
                  ОK
                          = 2'b11;
  always @ (posedge CLOCK) begin // State memory (with sync. reset)
   if (RESET==1) Sreg <= INIT; else Sreg <= Snext;</pre>
   LASTA \leq A:
  end
  always @ (A, B, LASTA, Sreg) begin // Next-state logic
    case (Sreg)
     INIT:
              Snext = LOOKING;
     LOOKING: if (A==LASTA)
                                 Snext = OK;
               else
                                     Snext = LOOKING;
     OK :
             if (B==1 || A==LASTA) Snext = OK;
               else
                                     Snext = LOOKING;
     default
                                     Snext = INIT;
   endcase
  end
  assig Z = (Sreg==OK) ? 1 : 0; // Output logic
endmodule
```

- Tab. 9
 - First always block creates both state memory and LASTA register
 - Second one creates next-state logic using our simplified approach
 - Z output is a simple combinational decode of OK state, so it is created using a continuous-assignment statement

- Ones-counting machine
 - Design a clocked synchronous state machine with two inputs, X and Y, and one output, Z
 - Output should be 1 if the number of 1 inputs on X and Y since reset is a multiple of 4, and 0 otherwise
- We can write a Verilog module for this problem directly, without constructing a state table

Table 10: Verilog module for a ones-counting machine.

```
module Vronescnt( CLOCK, RESET, X, Y, Z );
input CLOCK, RESET, X, Y;
output Z;
wire Z; // declared as wire for continuous assignment
reg [1:0] COUNT, NEXTCNT; // Current and next count, modulo 4
parameter [1:0] ZERO = 2'b00;
always @ (posedge CLOCK) // State memory (with sync. reset)
if (RESET==1) COUNT <= ZERO; else COUNT <= NEXTCNT;
always @ (COUNT, X, Y) // Counting logic
NEXTCNT = COUNT + {1'b0, X} + {1'b0, Y};
assign Z = (COUNT==ZERO) ? 1 : 0; // Output logic
endmodule
```

- In synthesis, counting logic in Tab. 10 does not necessarily yield a compact or speedy circuit
 - With a simple tool, it could yield two 2-bit adders connected in series
 - A good tool may be able to synthesize a more compact incrementer for each of the two additions
 - Another approach is to replace NEXTCNT always block with the one shown in Tab. 11
 - Formulating the choices in a case statement makes for a faster circuit, allowing the two adders or incrementers to operate in parallel
 - A multiplexer can be used to select one of outputs according to choices
 - ${\scriptstyle \bullet}\,$ To ensure that only one full adder is synthesized, we use code shown in Tab. 12

Table 11: Alternative Verilog counting logic for ones-counting machine.

always @ (COUNT	, X, Y) <mark>c</mark> a	ase ({X, Y})	<pre>// Counting logic</pre>
2'b01, 2'b10:	NEXTCNT =	= COUNT + 2'b01	;
2'b11:	NEXTCNT :	= COUNT + 2'b10	;
default	NEXTCNT :	= COUNT;	
endcase			

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Table 12:Fastest and smallest Verilog counting logic for ones-countingmachine.

```
always @ (COUNT, X, Y) begin // Counting logic
XY1s[0] = X ^ Y;
XY1s[1] = X & Y;
NEXTCNT = COUNT + XY1s;
end
```

- Tab. 12
 - A 2-bit reg variable XY1s is declared
 - The two equations create a half adder
 - Only one full adder is synthesized

- Combination-lock state machine
 - Design a clocked synchronous state machine with one input, X, and one output, UNLK
 UNLK output should be 1 if and only if X is 0 and the sequence of inputs received on X at preceding seven clock ticks was 0110111
 - Output of the machine at any time is completely determined by its inputs during current and preceding seven clock ticks
 - Thus, we use a "finite-memory" approach to design this machine
 - With this approach, we explicitly keep track of past seven inputs and then form output as a combinational function of these and current inputs
 - Tab. 13
 - This module keeps track of last seven values of X using a "shift register"
 - Bits are shifted left on each clock tick
 - Next-state logic is put into the same always block as state memory
 - XHISTORY is initialized to all 1s at reset so the user does not get benefit of a "free" 0 right after reset to begin combination pattern

Table 13: Verilog module for finite-memory design of combination-lock state machine.

References

JOHN F. WAKERLY, Digital Design: Principles and Practices (4th Edition), PRENTICE HALL, 2005.